



Pulsed Power Engineering Diagnostics

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Diagnostic Techniques and Considerations in Pulsed Power Systems



- Grounding
 - Proper grounding
 - Ground loops
- Voltage measurement
 - Voltage divider
 - Resistive
 - Capacitive
 - Balanced
 - Commercial voltage probes
- Current measurements
 - CVR
 - Rogowski
 - Self-integrating Rogowski



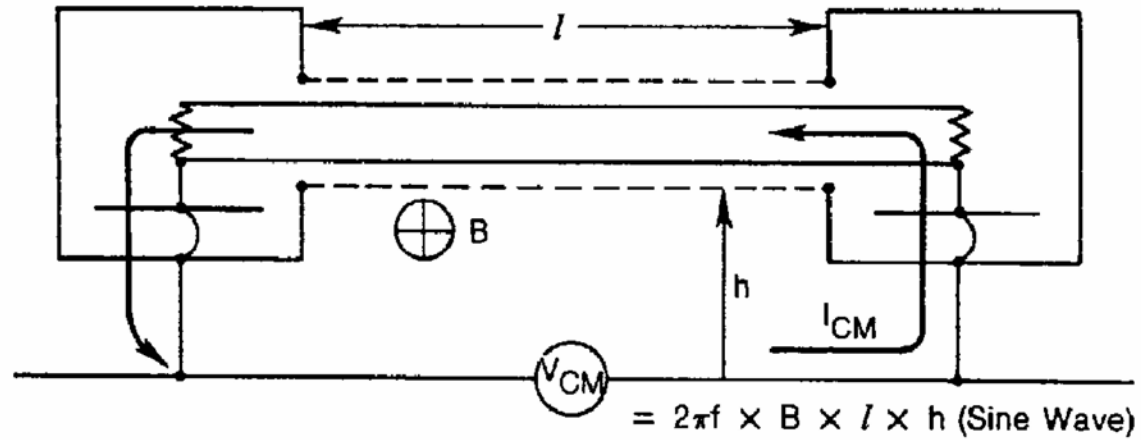
Grounding

- Proper grounding is the single most important factor in making accurate experimental measurements in pulsed power systems: design it in
- $\text{kA}/\mu\text{s} \times \text{nH} = \text{V}$, no two points in a high dI/dt system ever have the same potential which will induce “ground loop” currents
- Solid “earth” ground when possible
- “Single point” ground systems when possible
 - Almost anything with an AC plug has a ground lead
 - Safety requirements often result in additional grounds
- Use tri-axial cables instead of co-ax, outer shield can be non-current carrying connection required for grounding/bonding



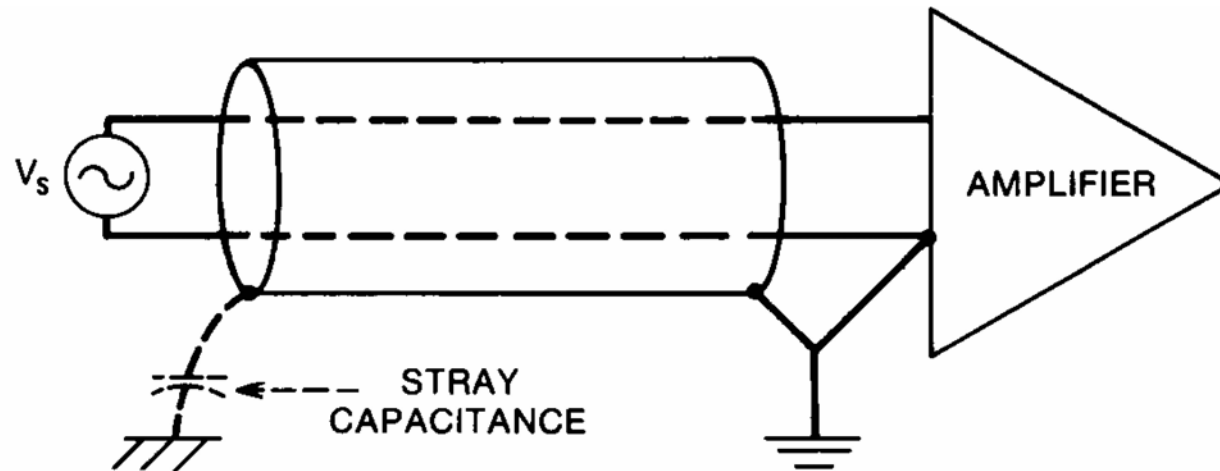
Ground Loop

DC coupled



$$\text{or } - \frac{\Delta B}{\Delta t} \times l \times h \text{ (Pulsed Field)}$$

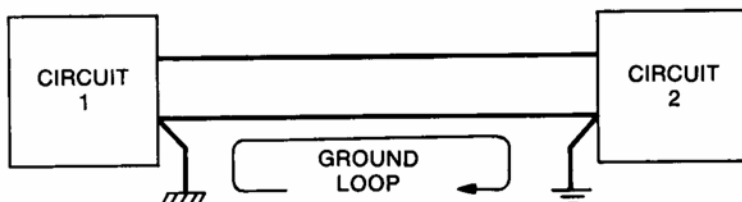
Capacitively coupled



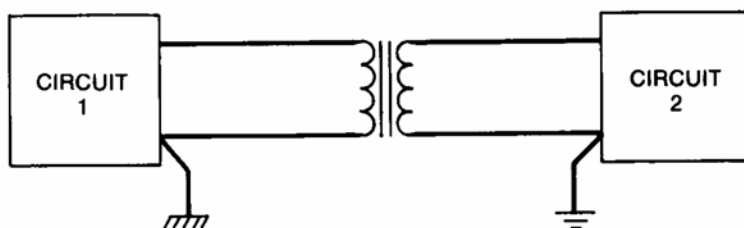


Isolation Techniques for Ground Loops

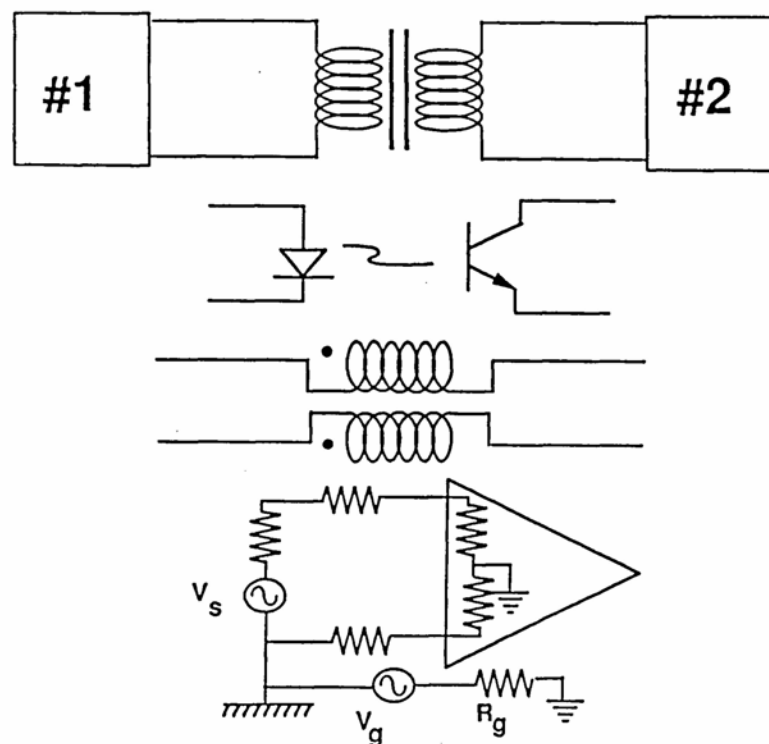
Ground loop from multiple-point grounding



Interrupting ground loop current flow using transformer isolation

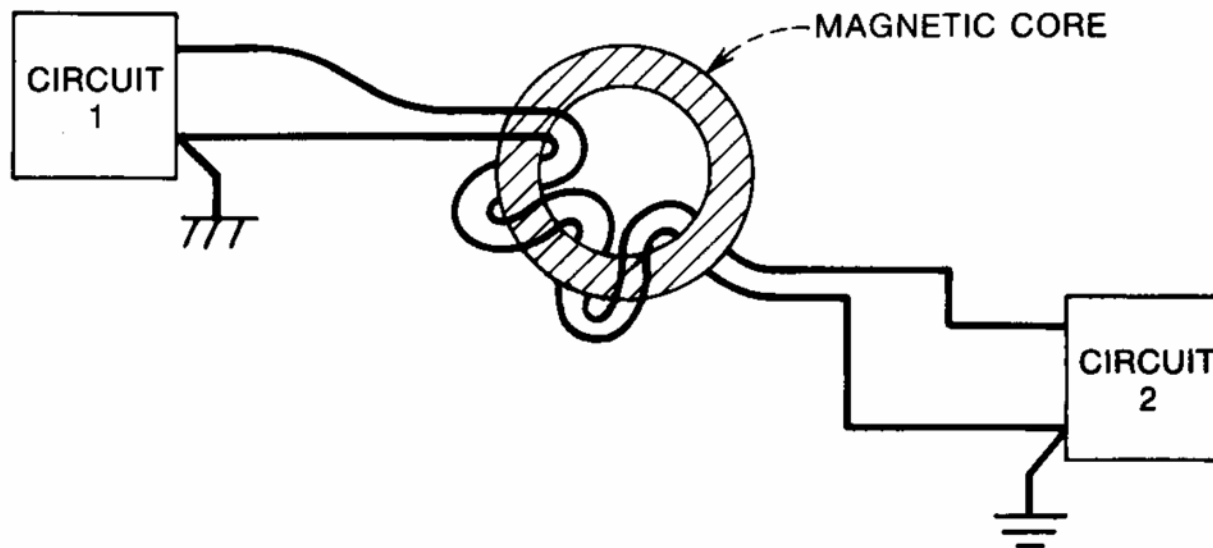


Additional isolation techniques





Common Mode Choke for Signal Cables





Measuring High Voltage

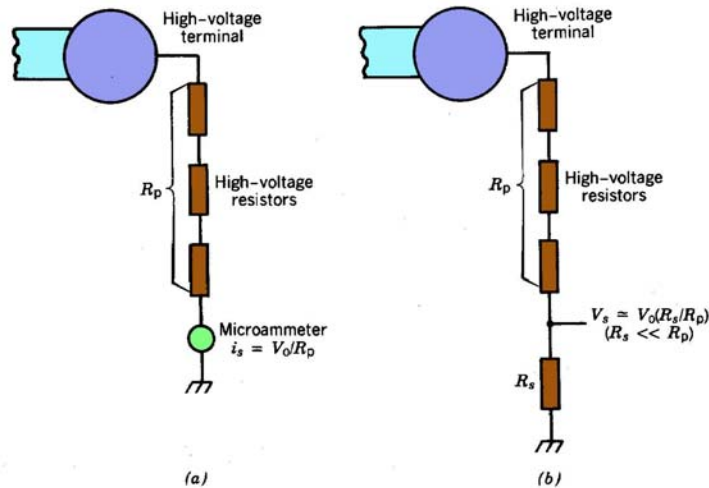


Figure 9.49 High-voltage measurements with resistor strings. (a) Resistive shunt. (b) Resistive divider.

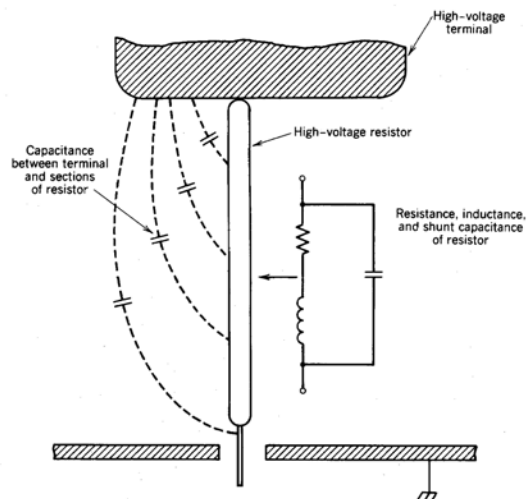


Figure 9.51 Effect of stray capacitance, shunt capacitance, and series inductance on resistive voltage divider.

- High voltage resistor strings are used to make HV measurements
 - Resistive shunts
 - Resistive dividers
- Parasitic effects (illustrated in Fig 9.51) can introduce waveform distortion at higher frequencies as illustrated in Fig 9.52
- Impact of parasitic elements is reduced as resistance of string is reduced, but dissipation and loss increases

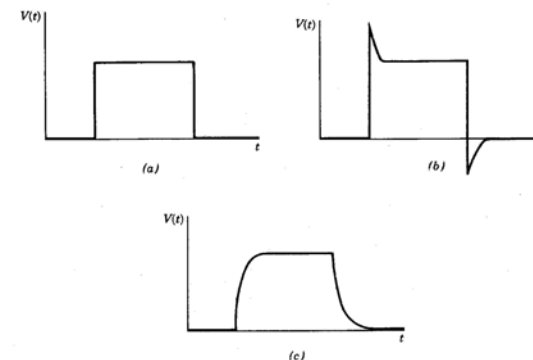


Figure 9.52 Waveforms from resistive voltage divider. (a) Ideal output voltage from a square-pulse input. (b) Output with significant shunt capacitance. (c) Output with significant series inductance.

High Frequency Voltage Dividers

- Most common alternatives
 - Capacitive divider
 - Balanced divider
 - Add capacitance to “swap” strays
 - Can be done with discrete components
 - Alt: physically divide resistive medium
 - Water
 - Thin film
 - Typical design of commercial HV probes
 - Inductive dividers used for dI/dt

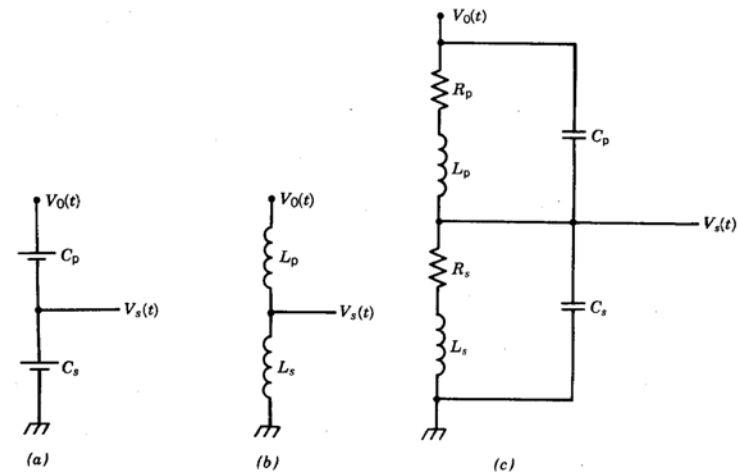


Figure 9.53 Voltage dividers. (a) Capacitive. (b) Inductive. (c) Balanced voltage divider.

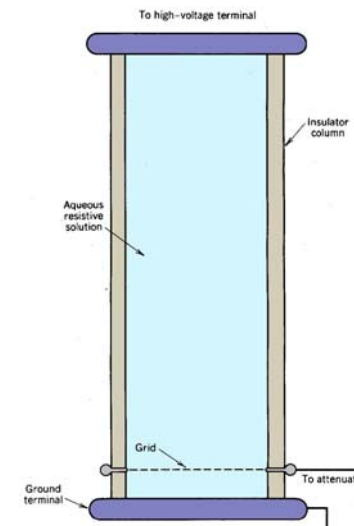


Figure 9.54 Balanced voltage divider with water solution resistor.



Scope Probes

- Balanced probes
 - Input impedance is frequency dependent
 - Scope impedance impacts response
- Bandwidth is limited
 - May be substantially less than rating, depending on ground connection
- HV versions require tuning to scope
- Pulsed power workhorses
 - P5100: 100X, 2.5 kV, 250 MHz
 - P6015: 1000X, 20 kV, 75 MHz
 - P5210 (differential): 5 kV, 50 MHz, 2 kV common-mode





Current Measurement

- Current viewing resistor
 - $V = IR$
- Time changing induced magnetic field, dB/dt
 - B-dot loop
 - $V = NA \text{ dB/dt}$
 - Coil of area, A, with N turns
 - $V = NAB/RC$
 - Passive RC integrator
 - Calibration difficult, function of source and loop
 - Location
 - Size
 - Orientation
 - Rogowski coil
 - Encloses current source
 - Eliminates location/orientation calibration factors

Rogowski Coil

- Usual “air core” approximation, diamagnetic field of loop is negligible
 - $B_i = B$
- $B(r) = \mu I / 2\pi r$
- $V = NA \, dB/dt$

$$= \mu A (N/2\pi r) \, dI/dt$$

$$= \mu A (N/\ell) \, dI/dt$$

$$= \mu A (N/\ell) I / RC \text{ (with RC integrator)}$$

ℓ is coil length

N/ℓ is number of turns/meter
- Can be built in the lab
 - Calibration challenges: accurately measuring A and N/ℓ
 - Signal attenuation from passive RC integrator yields small signals unless I very large or time constant short
- Commercially available

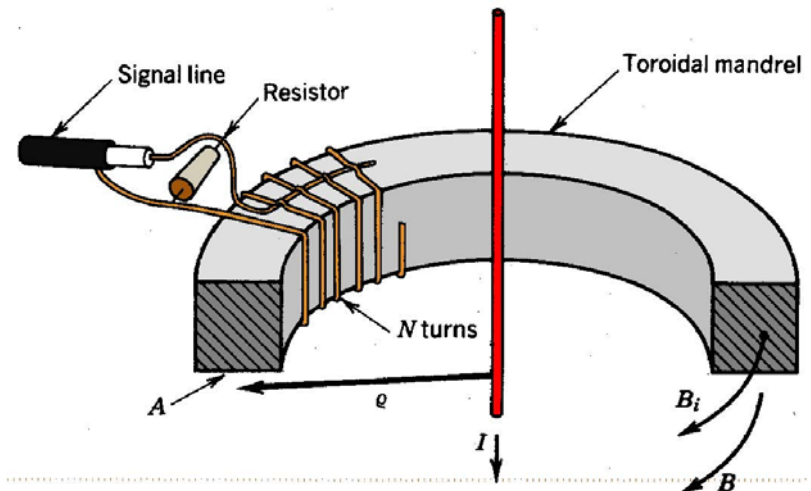


Figure 9.58 Rogowski loop—self-integrating current probe.





Self-integrating Rogowski

- More rigorously, the field B_i , in Fig 9.58

$$B_i = B - \mu i (N/2\pi\rho)$$

where i is the current flowing in the coil

$$i = NA (dB_i/dt)/R$$

- Combining the above and solving for B

$$B = B_i + (dB_i/dt) (\mu N^2 A/2\pi\rho R)$$

$$= B_i + (dB_i/dt) (L/R) \text{ inserting the identity for a solenoid inductor}$$

- When the time constant L/R is large compared to the time scale of current variations: $(d/dt) (L/R) \gg 1$, then the left term above can be neglected and:

$$B \approx (dB_i/dt) (\mu N^2 A/2\pi\rho R)$$

- Recognizing $B = \mu I/2\pi\rho$ and solving for dB_i/dt as a function of coil current

$$i = I/N$$

- Typically, L is made large by using a ferrite core
- Commercial current transformer

